

CBCS SCHEME

USN

2	V	0	2	1	C	C	0	0	8
---	---	---	---	---	---	---	---	---	---

21EC33

Third Semester B.E. Degree Examination, Jan./Feb. 2023 Basic Signal Processing

Time: 3 hrs.

Max. Marks: 100

Note: Answer any FIVE full questions, choosing ONE full question from each module.

Module-1

1.
 - a. Explain vector spaces and its necessary axioms. And also explain four fundamental subspaces with example. (08 Marks)
 - b. Write the vector $V = (1, 3, 9)$ as a linear combination of the vectors $u_1 = (2, 1, 3)$, $u_2 = (1, -1, 1)$ and $u_3 = (3, 1, 5)$ and thereby show that the system is consistent. (08 Marks)
 - c. Let $f: V_1(\mathbb{R}) \rightarrow V_2(\mathbb{R})$ be a mapping $f(x) = (3x, 5x)$ show that 'f' is linear transformation. (04 Marks)

OR

2.
 - a. Let 'w' be the subspace of \mathbb{R}^5 spanned by
 $x_1 = (1, 2, -1, 3, 4)$, $x_2 = (2, 4, -2, 6, 8)$, $x_3 = (1, 3, 2, 2, 6)$,
 $x_4 = (1, 4, 5, 1, 8)$, $x_5 = (2, 7, 3, 3, 9)$.
 Find a subset of vectors which forms a basis of 'w'. (06 Marks)
 - b. Solve $Ax = b$ by least square and find $P = Ax$ if
 $A = \begin{bmatrix} 4 & 0 \\ 0 & 2 \\ 1 & 1 \end{bmatrix}_{3 \times 2}$ and $b = \begin{bmatrix} 2 \\ 0 \\ 11 \end{bmatrix}_{3 \times 1}$. Also, write a program to solve linear equation $Ax = b$. (07 Marks)
 - c. Apply Gram - Schmidt process to the vectors $V_1(1, 1, 1)$, $V_2(1, -1, 2)$, $V_3(2, 1, 2)$ to obtain an orthonormal basis for $V_3(\mathbb{R})$ with standard inner product and thereby write a program for Gram - Schmidt process. (07 Marks)

Module-2

3.
 - a. If $A = \begin{bmatrix} 4 & 2 & -2 \\ -5 & 3 & 2 \\ -2 & 4 & 1 \end{bmatrix}$ find Eigen values and corresponding Eigen vector for matrix 'A' and diagonalize the matrix. (10 Marks)
 - b. If $A = \begin{bmatrix} 3 & -1 & 1 \\ -1 & 5 & -1 \\ 1 & -1 & 3 \end{bmatrix}$. Show that matrix 'A' is positive definite matrix using the following approaches :
 i) By finding its Eigen value
 ii) By finding its pivots. (10 Marks)

Important Note: 1. On completing your answers, compulsorily draw diagonal cross lines on the remaining blank pages. 2. Any revealing of identification, appeal to evaluator and/or equations written eg. 42 → 4 = 50, will be treated as malpractice.

OR

- 4 a. Compute $A^T A$ and AA^T , find Eigen values and Eigen vectors, if $A = \begin{bmatrix} 1 & 1 \\ 0 & 1 \\ -1 & 1 \end{bmatrix}_{3 \times 2}$, thereby multiply U & V^T to recover matrix 'A'. Also write a program to find SVD. (12 Marks)

- b. Diagonalize the matrix A, if $A = \begin{bmatrix} -2 & 2 & -3 \\ 2 & 1 & -6 \\ -1 & -2 & 0 \end{bmatrix}$ by finding its eigen value and eigen vector. (08 Marks)

Module-3

- 5 a. Define signal and system and also explain basic discrete elementary signals with neat sketch and expressions. (04 Marks)
- b. A discrete time signal $x(n]$ is shown below Fig.Q5(b).

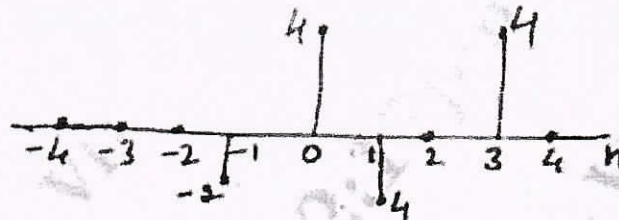


Fig.Q5(b)

Sketch :

- i) $2x(n-2)$
 ii) $3-x(n)$
 iii) $2x(-n)-4$.

(08 Marks)

- c. Sketch : $x(n) = \begin{cases} 1; & -1 \leq n \leq 3 \\ \frac{1}{2}; & n = 4 \\ 0; & \text{otherwise} \end{cases}$ and $y(n) = \begin{cases} \frac{1}{2}n; & |n| \leq 4 \\ 0; & \text{otherwise} \end{cases}$

Also sketch $x(n+2)y(1-2n)$.

(08 Marks)

OR

- 6 a. For the following discrete time systems, determine whether the system is linear, time invariance, memoryless, causal and stable :

i) $y(n) = 2x(n) + \frac{1}{x(n-2)}$

ii) $y(n) = \ln(3 + |x(n)|)$

iii) $y(n) = \cos x(n)$

iv) $y(n) = r^n x(n); r > 1$.

(16 Marks)

- b. Write a program to generate exponential and triangular waveforms.

(04 Marks)

Module-4

- 7 a. Compute the discrete time convolution for the sequences $x_1(n)$ and $x_2(n)$ given below
 $x_1(n) = \alpha^n u(n)$; $x_2(n) = \beta^n u(n)$. (08 Marks)
- b. Consider the input signal $x(n)$ and the impulse response $h(n)$ given below:

$$x(n) = \begin{cases} 1, & 0 \leq n \leq 4 \\ 0, & \text{otherwise} \end{cases} \quad \text{and} \quad h(n) = \begin{cases} \alpha^n, & 0 \leq n \leq 6, \alpha > 1 \\ 0, & \text{otherwise} \end{cases}$$

compute the output signal $y(n)$. (12 Marks)

OR

- 8 a. The following are the impulse responses of discrete time LTI systems. Determine whether each system is memoryless, causal and stable:

i) $h(n) = e^{-n} \cos(n) \cdot u(n)$

ii) $h(n) = (0.99)^n u(n+3)$

iii) $h(n) = u\left(\frac{1}{2}\right)^n u(n)$. (10 Marks)

- b. Evaluate the step response of LTI system represented by the impulse response

$$h(n) = (-1)^n \{u(n+2) - u(n-3)\}.$$

Also write a program to compute the step response from the given impulse response.

(10 Marks)

Module-5

- 9 a. Define Z-transform. Explain the properties of ROC. (06 Marks)

b. Let $x(n) = \left(\frac{1}{2}\right)^n u(n)$.

i) Sketch $x(n)$

ii) Find $X(z)$ and sketch pole zero plot and ROC (08 Marks)

c. Find the Z-transform of $x(n) = \left(\frac{1}{2}\right)^n u(n) + \left(\frac{1}{3}\right)^n u(n)$ (06 Marks)

OR

- 10 a. Explain the properties of Z-transform with proof:

i) Convolution

ii) Initial value theorem

iii) Final value theorem. (08 Marks)

- b. Determine the describe time sequence $x(n)$ of the sequence using partial fraction expression:

$$X(z) = \frac{-1 + 5z^{-1}}{1 - \frac{1}{2}z^{-1} + \frac{1}{2}z^{-2}}, \text{ ROC: } |z| > 1$$

(08 Marks)

- c. Write a program to find Z-transform of the sequence. (04 Marks)

KLS Vishwanathrao Deshpande Institute of
Technology


Department of Electronics and Communication Engg


Subject: Basic Signal Processing Jan/Feb 2023
Scheme and Solution.


Subject code: 21 EC33

Sem : 3

Staff: Prof. Plasin. F. Dias.


Staff Incharge
Prof. Plasin. F. Dias.


HOD
Dr. Mabendra M. Desai
Head of the Department
Dept. of Electronic & Communication Engg.
KLS V.D.I.T. HALIYAL (U.K.)


Dean
Academics
Dr. S. Deshpande
Dean, Academics
KLS VDIT, HALIYAL

Module 1

1 a Vector Space

It is a non empty set V of objects called vectors on which are two defined two operations called addition and multiplication by scalars.

Axioms:

1. Closed under Addition
2. Commutative under addition
3. Associative under addition
4. Additive Identity : There exist 0 in V , such that $u + 0 = u$, for all u in V ,
5. Additive Inverse: for all u in V , there exists $-u$ in V such that $u + (-u) = 0$.
6. closed under scalar multiplication
7. Multiplicative Identity
8. Associative under scalar multiplication
9. Distributive under scalar multiplication

Fundamental Subspaces:

1. column space of A , is denoted by $C(A)$. Its dimension is rank r .
2. Null space of A denoted by $N(A)$, Its dimension is $n-r$,
3. Row space of A is column space of A^T . It is $C(A^T)$ and is spanned by rows of A . Its dimension is also r .
4. Left null space of A is null space of A^T . It contains all vectors y such that $A^T y = 0$, and it is written $N(A^T)$. Its dimension is $m-r$.

1 b.

$$au_1 + bu_2 + cu_3 = v$$

$$a \begin{bmatrix} 2 \\ 1 \\ 3 \end{bmatrix} + b \begin{bmatrix} 1 \\ -1 \\ 1 \end{bmatrix} + c \begin{bmatrix} 3 \\ 1 \\ 5 \end{bmatrix} = \begin{bmatrix} 1 \\ 3 \\ 9 \end{bmatrix}$$

Equations are,

$$2a + b + 3c = 1$$

$$a - b + c = 3$$

$$3a + b + 5c = 9$$

Solving above equations,

$$a = -12$$

$$b = -5$$

$$c = 10$$

Since components of u_1 are coefficients of a ,

Components of u_2 are coefficients of b

Components of u_3 are coefficients of c

The given vector v is a linear combination of u_1, u_2 & u_3

from above equations, it is realized, the system is consistent and it has at least one solution.

c. Given $f(x) = (3x, 5x)$

$$\begin{aligned} \text{i) } f(x_1 + x_2) &= \{ 3(x_1 + x_2), 5(x_1 + x_2) \} \\ &= (3x_1 + 3x_2, 5x_1 + 5x_2) \\ &= (3x_1, 5x_1) + (3x_2, 5x_2) \end{aligned}$$

$$f(x_1 + x_2) = f(x_1) + f(x_2)$$

$$\begin{aligned} \text{(ii) } f(ax) &= \{ 3(ax), 5(ax) \} \\ &= (3ax, 5ax) \\ &= a(3x, 5x) \end{aligned}$$

$$f(ax) = a(f(x)), \quad f \text{ is linear transformation of } x.$$

2 a.

$$\alpha_1 = (1, 2, -1, 3, 4)$$

$$\alpha_2 = (2, 4, -2, 6, 8)$$

$$\alpha_3 = (1, 3, 2, 2, 6)$$

$$\alpha_4 = (1, 4, 5, 1, 8)$$

$$\alpha_5 = (2, 7, 3, 3, 9)$$

$$\text{Let } W = \begin{bmatrix} 1 & 2 & -1 & 3 & 4 \\ 2 & 4 & -2 & 6 & 8 \\ 1 & 3 & 2 & 2 & 6 \\ 1 & 4 & 5 & 1 & 8 \\ 2 & 7 & 3 & 3 & 9 \end{bmatrix}$$

$$R_2 \rightarrow R_2/2 - R_1$$

$$R_3 \rightarrow R_3 - R_1$$

$$W = \begin{bmatrix} 1 & 2 & -1 & 3 & 4 \\ 0 & 0 & 0 & 0 & 0 \\ 0 & 1 & 3 & -1 & 2 \\ 1 & 4 & 5 & 1 & 8 \\ 2 & 7 & 3 & 3 & 9 \end{bmatrix}$$

$$R_4 \rightarrow R_4 - R_1, \quad R_5 \rightarrow R_5 - 2R_1$$

$$W = \begin{bmatrix} 1 & 2 & -1 & 3 & 4 \\ 0 & 0 & 0 & 0 & 0 \\ 0 & 1 & 3 & -1 & 2 \\ 0 & 2 & 6 & -2 & 4 \\ 0 & 3 & 5 & -3 & 1 \end{bmatrix}$$

$$R_4 \rightarrow R_4 - 2R_3, \quad R_2 \leftrightarrow R_3$$

$$W = \begin{bmatrix} 1 & 0 & -7 & 5 & 0 \\ 0 & 1 & 3 & -1 & 2 \\ 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 \\ 0 & 3 & 5 & -3 & 1 \end{bmatrix}$$

$$R_5 - 3R_2 \leftarrow R_5$$

$$W = \begin{bmatrix} 1 & 0 & -7 & 5 & 0 \\ 0 & 1 & 3 & -1 & 2 \\ 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & -4 & 0 & -5 \end{bmatrix}$$

X

$$R_3 \leftrightarrow R_5$$

$$W = \begin{bmatrix} 1 & 0 & -7 & 5 & 0 \\ 0 & 1 & 3 & -1 & 2 \\ 0 & 0 & -4 & 0 & -5 \\ 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 \end{bmatrix}$$

The above matrix is row reduced, hence basis of W will be

$$\left\{ (1, 0, -7, 5, 0), (0, 1, 3, -1, 2), (0, 0, -4, 0, -5) \right\}$$

2 b.

$$A = \begin{bmatrix} 4 & 0 \\ 0 & 2 \\ 1 & 1 \end{bmatrix} \quad A^T = \begin{bmatrix} 4 & 0 & 1 \\ 0 & 2 & 1 \end{bmatrix}$$

$$A \vec{x} = b$$

$$A^T A \vec{x} = A^T b$$

$$A^T A = \begin{bmatrix} 17 & 1 \\ 1 & 5 \end{bmatrix}$$

$$A^T b = \begin{bmatrix} 4 & 0 & 1 \\ 0 & 2 & 1 \end{bmatrix} \begin{bmatrix} 2 \\ 0 \\ 11 \end{bmatrix} = \begin{bmatrix} 19 \\ 11 \end{bmatrix}$$

$$17x_1 + 1x_2 = 19$$

$$x_1 + 5x_2 = 11$$

$$x_1 = 1, \quad x_2 = 2$$

$$\vec{x} = \begin{bmatrix} 1 \\ 2 \end{bmatrix}$$

$$P = A \vec{x}$$

$$P = \begin{bmatrix} 4 & 0 \\ 0 & 2 \\ 1 & 1 \end{bmatrix} \begin{bmatrix} 1 \\ 2 \end{bmatrix}$$

Q.E.D.

$$P = \begin{bmatrix} 4 \\ 4 \\ 3 \end{bmatrix}$$

from sympy import symbols, Eq, solve
 $x, y = \text{symbols}('x, y')$

defining equations

$$\text{eq1} = \text{Eq}((x+y), 1)$$

print ("Equation 1:")

print (eq1)

$$\text{eq2} = \text{Eq}((x-y), 1)$$

print ("Equation 2:")

print (eq2)

solving the equations

print ("Values of 2 unknown variable
 are as follows:")

print (solve((eq1, eq2), (x, y)))

$$v_1 = \begin{bmatrix} 1 \\ 1 \\ 1 \end{bmatrix},$$

$$v_2 = \begin{bmatrix} 1 \\ -1 \\ 2 \end{bmatrix}$$

$$v_3 = \begin{bmatrix} 2 \\ 1 \\ 2 \end{bmatrix}$$

$$u_1 = \frac{v_1}{\|v_1\|} = \frac{(1, 1, 1)}{\sqrt{3}}$$

$$u_1 = \begin{bmatrix} 1/\sqrt{3} \\ 1/\sqrt{3} \\ 1/\sqrt{3} \end{bmatrix}$$

$$u_2 = \frac{u_2'}{\|u_2'\|}$$

$$u_2' = v_2 - \text{proj}_{u_1}(v_2)$$

$$= v_2 - (u_1 \cdot v_2) u_1$$

$$= \begin{bmatrix} 1 \\ -1 \\ 2 \end{bmatrix} - \left(\frac{1}{\sqrt{3}} \cdot 1 - \frac{1}{\sqrt{3}} \cdot 1 + \frac{1}{\sqrt{3}} \cdot 2 \right) \begin{bmatrix} \frac{1}{\sqrt{3}} \\ \frac{1}{\sqrt{3}} \\ \frac{1}{\sqrt{3}} \end{bmatrix}$$

$$= \begin{bmatrix} 1 \\ -1 \\ 2 \end{bmatrix} - \begin{bmatrix} \frac{2}{3} \\ \frac{2}{3} \\ \frac{2}{3} \end{bmatrix}$$

$$u_2' = \begin{bmatrix} \frac{1}{3} \\ -\frac{5}{3} \\ \frac{4}{3} \end{bmatrix}$$

$$\|u_2'\| = \sqrt{\frac{14}{3}} = \frac{\sqrt{42}}{3}$$

$$u_2 = \begin{bmatrix} \frac{1}{\sqrt{42}} \\ -\frac{5}{\sqrt{42}} \\ \frac{4}{\sqrt{42}} \end{bmatrix}$$

$$u_3 = \frac{u_3'}{\|u_3'\|}$$

$$u_3' = v_3 - \text{proj}_{u_1}(v_3) - \text{proj}_{u_2}(v_3)$$

$$= v_3 - (u_1 \cdot v_3) u_1 - (u_2 \cdot v_3) u_2$$

$$= \begin{bmatrix} 2 \\ 1 \\ 2 \end{bmatrix} - \left(\frac{2}{\sqrt{3}} + \frac{1}{\sqrt{3}} + \frac{2}{\sqrt{3}} \right) \begin{bmatrix} \frac{1}{\sqrt{3}} \\ \frac{1}{\sqrt{3}} \\ \frac{1}{\sqrt{3}} \end{bmatrix} -$$

$$\left(\frac{2}{\sqrt{42}} - \frac{5}{\sqrt{42}} + \frac{8}{\sqrt{42}} \right) \begin{bmatrix} \frac{1}{\sqrt{42}} \\ -\frac{5}{\sqrt{42}} \\ \frac{4}{\sqrt{42}} \end{bmatrix}$$

$$= \begin{bmatrix} 2 \\ 1 \\ 2 \end{bmatrix} - \frac{5}{\sqrt{3}} \begin{bmatrix} \frac{1}{\sqrt{3}} \\ \frac{1}{\sqrt{3}} \\ \frac{1}{\sqrt{3}} \end{bmatrix} - \frac{5}{\sqrt{42}} \begin{bmatrix} \frac{1}{\sqrt{42}} \\ -\frac{5}{\sqrt{42}} \\ \frac{4}{\sqrt{42}} \end{bmatrix}$$

$$= \begin{bmatrix} 2 \\ 1 \\ 2 \end{bmatrix} - \begin{bmatrix} 5/3 \\ 5/3 \\ 5/3 \end{bmatrix} - \begin{bmatrix} 5/42 \\ -25/42 \\ 20/42 \end{bmatrix}$$

$$u_3' = \begin{bmatrix} 3/14 \\ -1/14 \\ -1/7 \end{bmatrix}$$

$$\|u_3'\| = 1/\sqrt{14}$$

$$u_3 = \begin{bmatrix} 3/\sqrt{14} \\ -1/\sqrt{14} \\ -\sqrt{14}/7 \end{bmatrix}$$

$$u_1 = \begin{bmatrix} 1/\sqrt{3} \\ 1/\sqrt{3} \\ 1/\sqrt{3} \end{bmatrix}$$

$$u_2 = \begin{bmatrix} 1/\sqrt{42} \\ -5/\sqrt{42} \\ 4/\sqrt{42} \end{bmatrix}$$

$$u_3 = \begin{bmatrix} 3/\sqrt{14} \\ -1/\sqrt{14} \\ -\sqrt{14}/7 \end{bmatrix}$$

```
import numpy as np
```

```
def gram-schmidt(A):
```

```
    """ orthogonalize a set of vectors """
```

```
    n = A.shape[1]
```

```
    for j in range(n):
```

```
        for k in range(j):
```

```
            A[:,j] = np.dot(A[:,k], A[:,j]) * A[:,k]
```

```
            A[:,j] = A[:,j] / np.linalg.norm(A[:,j])
```

```
    return A
```

```
if name == 'main':
```

```
A = np.array([[1.0, 1.0, 2.0], [1.0, -1.0, 1.0], [1.0, 2.0, 2.0]])
```

```
print(gram-schmidt(A))
```

Module-2

3 a.

$$A = \begin{bmatrix} 4 & 2 & -2 \\ -5 & 3 & 2 \\ -2 & 4 & 1 \end{bmatrix}$$

$$A - \lambda I = \begin{bmatrix} 4 & 2 & -2 \\ -5 & 3 & 2 \\ -2 & 4 & 1 \end{bmatrix} - \lambda \begin{bmatrix} 1 & 0 & 0 \\ 0 & 1 & 0 \\ 0 & 0 & 1 \end{bmatrix}$$

$$A - \lambda I = \begin{bmatrix} 4-\lambda & 2 & -2 \\ -5 & 3-\lambda & 2 \\ -2 & 4 & 1-\lambda \end{bmatrix}$$

$$|A - \lambda I| = \lambda^3 - S_1 \lambda^2 + S_2 \lambda - S_3$$

$$S_1 = 4 + 3 + 1 = 8$$

$$S_2 = (3-8) + (4-4) + (12+10) = 17$$

$$S_3 = 4(3-8) - 2(-5+4) - 2(-20+6) \\ = 4(-5) - 2(-1) - 2(-14)$$

$$S_3 = 10$$

$$|A - \lambda I| = \lambda^3 - 8\lambda^2 + 17\lambda - 10 = 0$$

By trial & error method $\lambda = 5$ is a root of a above equation,

$$5 \left| \begin{array}{cccc} 1 & -8 & 17 & -10 \\ 0 & 5 & -15 & 10 \\ \hline 1 & -3 & 2 & 0 \end{array} \right.$$

$$\therefore \lambda^2 - 3\lambda + 2 = 0$$

$$\lambda^2 - 2\lambda - \lambda + 2 = 0$$

$$\lambda(\lambda - 2) - (\lambda - 2) = 0$$

$$(\lambda - 1)(\lambda - 2) = 0$$

$$\lambda = 1, \lambda = 2$$

~~2~~

Hence the roots are $\lambda_1 = 5, \lambda_2 = 2, \lambda_3 = 1$

for $\lambda_1 = 5$

$$A - \lambda_1 I = A - 5I = \begin{bmatrix} -1 & 2 & -2 \\ -5 & -2 & 2 \\ -2 & 4 & -4 \end{bmatrix}$$

$$[A - \lambda_1 I] \vec{x} = 0$$

$$\begin{bmatrix} -1 & 2 & -2 \\ -5 & -2 & 2 \\ -2 & 4 & -4 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ x_3 \end{bmatrix} = 0$$

$$-x_1 + 2x_2 - 2x_3 = 0$$

$$-5x_1 - 2x_2 + 2x_3 = 0$$

By Cramer's Rule,

$$\frac{x_1}{\begin{vmatrix} +2 & -2 \\ -2 & 2 \end{vmatrix}} = \frac{-x_2}{\begin{vmatrix} -1 & -2 \\ -5 & 2 \end{vmatrix}} = \frac{x_3}{\begin{vmatrix} -1 & 2 \\ -5 & -2 \end{vmatrix}}$$

$$\frac{x_1}{0} = \frac{-x_2}{-12} = \frac{x_3}{12}$$

$$\frac{x_1}{0} = \frac{+x_2}{1} = \frac{x_3}{1}$$

$$\vec{x} = \begin{bmatrix} x_1 \\ x_2 \\ x_3 \end{bmatrix} = \begin{bmatrix} 0 \\ 1 \\ 1 \end{bmatrix}$$

for $\lambda_2 = 2$

$$[A - \lambda_2 I] = A - 2I = \begin{bmatrix} 2 & -2 & -2 \\ -5 & 1 & 2 \\ -2 & 4 & -1 \end{bmatrix}$$

$$[A - 2I] \vec{y} = 0$$

$$\begin{bmatrix} 2 & -2 & -2 \\ -5 & 1 & 2 \\ -2 & 4 & -1 \end{bmatrix} \begin{bmatrix} y_1 \\ y_2 \\ y_3 \end{bmatrix} = 0$$

$$2y_1 + 2y_2 - 2y_3 = 0$$

$$-5y_1 + y_2 + 2y_3 = 0$$

$$\frac{y_1}{\begin{vmatrix} 2 & -2 \\ 1 & 2 \end{vmatrix}} = \frac{-y_2}{\begin{vmatrix} 2 & -2 \\ -5 & 2 \end{vmatrix}} = \frac{y_3}{\begin{vmatrix} 2 & 2 \\ -5 & 1 \end{vmatrix}}$$

$$\frac{y_1}{6} = \frac{-y_2}{-6} = \frac{y_3}{12}$$

$$\frac{y_1}{1} = \frac{y_2}{1} = \frac{y_3}{2}$$

$$\vec{y} = \begin{bmatrix} y_1 \\ y_2 \\ y_3 \end{bmatrix} = \begin{bmatrix} 1 \\ 1 \\ 2 \end{bmatrix}$$

for $\lambda_3 = 1$

$$A - \lambda_3 I = A - I = \begin{bmatrix} 3 & 2 & -2 \\ -5 & 2 & 2 \\ -2 & 4 & 0 \end{bmatrix}$$

$$[A - I] \vec{z} = 0$$

$$\begin{bmatrix} 3 & 2 & -2 \\ -5 & 2 & 2 \\ -2 & 4 & 0 \end{bmatrix} \begin{bmatrix} z_1 \\ z_2 \\ z_3 \end{bmatrix} = 0$$

$$3z_1 + 2z_2 - 2z_3 = 0$$

$$-5z_1 + 2z_2 + 2z_3 = 0$$

$$\frac{z_1}{\begin{vmatrix} 2 & -2 \\ 2 & 2 \end{vmatrix}} = \frac{-z_2}{\begin{vmatrix} 3 & -2 \\ -5 & 2 \end{vmatrix}} = \frac{z_3}{\begin{vmatrix} 3 & 2 \\ -5 & 2 \end{vmatrix}}$$

$$\frac{z_1}{8} = \frac{-z_2}{-4} = \frac{z_3}{16}$$

$$\frac{z_1}{2} = \frac{z_2}{1} = \frac{z_3}{4}$$

$$\vec{z} = \begin{bmatrix} z_1 \\ z_2 \\ z_3 \end{bmatrix} = \begin{bmatrix} 2 \\ 1 \\ 4 \end{bmatrix}$$

$$S_3 = 3(15+1) + (-3+1) + (-1-5)$$

$$= 48 - 2 - 6$$

$$S_3 = 40.$$

$$|A - \lambda I| = \lambda^3 - 11\lambda^2 + 38\lambda - 40 = 0$$

By trial & error method,

$\lambda = 2$ is a factor for above equation.

$$\begin{array}{c} \div \\ \hline \begin{array}{cccc} 1 & -11 & 38 & -40 \\ 0 & 2 & -18 & 40 \\ \hline 1 & -9 & 20 & 0 \end{array} \end{array}$$

$$\lambda^2 - 9\lambda + 20 = 0$$

$$\lambda^2 - 5\lambda - 4\lambda + 20 = 0$$

$$\lambda(\lambda - 5) - 4(\lambda - 5) = 0$$

$$(\lambda - 4)(\lambda - 5) = 0$$

$$\lambda = 4, \lambda = 5$$

The roots are $\lambda = 2, 4, 5$

Since the eigen values are greater than zero, the matrix is said to be positive definite.

$$A = \begin{bmatrix} 3 & -1 & 1 \\ -1 & 5 & -1 \\ 1 & -1 & 3 \end{bmatrix}$$

$$R_2 \rightarrow 3R_2 + R_1 \quad R_3 \rightarrow 3R_3 - R_1$$

$$A = \begin{bmatrix} 3 & -1 & 1 \\ 0 & 14 & -2 \\ 0 & -2 & 8 \end{bmatrix}$$

$$R_2 \rightarrow R_2/2 \quad , \quad R_3 \rightarrow R_3/2$$

$$A = \begin{bmatrix} 3 & -1 & 1 \\ 0 & 7 & -1 \\ 0 & -1 & 4 \end{bmatrix}$$

$$P = \begin{bmatrix} 0 & 1 & 2 \\ 1 & 1 & 1 \\ 1 & 2 & 4 \end{bmatrix}$$

$$D = \begin{bmatrix} 5 & 0 & 0 \\ 0 & 2 & 0 \\ 0 & 0 & 1 \end{bmatrix}$$

$$P^{-1} = \begin{bmatrix} -2 & 0 & 1 \\ 3 & 2 & -2 \\ -1 & -1 & 1 \end{bmatrix}$$

$$A = P D P^{-1}$$

$$\begin{bmatrix} 4 & 2 & -2 \\ -5 & 3 & 2 \\ -2 & 4 & 1 \end{bmatrix} = \begin{bmatrix} 0 & 1 & 2 \\ 1 & 1 & 1 \\ 1 & 2 & 4 \end{bmatrix} \begin{bmatrix} 5 & 0 & 0 \\ 0 & 2 & 0 \\ 0 & 0 & 1 \end{bmatrix}$$

$$\begin{bmatrix} 4 & 2 & -2 \\ -5 & 3 & 2 \\ -2 & 4 & 1 \end{bmatrix} = \begin{bmatrix} 4 & 2 & -2 \\ -5 & 3 & 2 \\ -2 & 4 & 1 \end{bmatrix}$$

which is diagonalized matrix

3 b.

$$A = \begin{bmatrix} 3 & -1 & 1 \\ -1 & 5 & -1 \\ 1 & 1 & 3 \end{bmatrix}$$

$$A - \lambda I = \begin{bmatrix} 3 & -1 & 1 \\ -1 & 5 & -1 \\ 1 & 1 & 3 \end{bmatrix} - \lambda \begin{bmatrix} 1 & 0 & 0 \\ 0 & 1 & 0 \\ 0 & 0 & 1 \end{bmatrix}$$

$$A - \lambda I = \begin{bmatrix} 3-\lambda & -1 & 1 \\ -1 & 5-\lambda & -1 \\ 1 & 1 & 3-\lambda \end{bmatrix}$$

$$|A - \lambda I| = \lambda^3 - S_1 \lambda^2 + S_2 \lambda - S_3 = 0$$

$$S_1 = 3 + 5 + 3 = 11$$

$$S_2 = (15+1) + (9-1) + (15-1) = 38$$

~~2~~

$$R_3 \rightarrow 7R_3 + R_2$$

$$A = \begin{bmatrix} 3 & -1 & 1 \\ 0 & 7 & -1 \\ 0 & 0 & 27 \end{bmatrix}$$

The pivots are 3, 7, 27, since pivots are greater than zero matrix is positive definite.

4 a)

$$A = \begin{bmatrix} 1 & 1 \\ 0 & 1 \\ -1 & 1 \end{bmatrix} \quad A^T = \begin{bmatrix} 1 & 0 & -1 \\ 1 & 1 & 1 \end{bmatrix}$$

$$A^T A = \begin{bmatrix} 1 & 0 & -1 \\ 1 & 1 & 1 \end{bmatrix} \begin{bmatrix} 1 & 1 \\ 0 & 1 \\ -1 & 1 \end{bmatrix} = \begin{bmatrix} 2 & 0 \\ 0 & 3 \end{bmatrix}$$

$$A A^T = \begin{bmatrix} 1 & 1 \\ 0 & 1 \\ -1 & 1 \end{bmatrix} \begin{bmatrix} 1 & 0 & -1 \\ 1 & 1 & 1 \end{bmatrix}$$

$$A A^T = \begin{bmatrix} 2 & 1 & 0 \\ 1 & 1 & 1 \\ 0 & 1 & 2 \end{bmatrix}$$

$$A^T A - \lambda I = \begin{bmatrix} 2 & 0 \\ 0 & 3 \end{bmatrix} - \begin{bmatrix} \lambda & 0 \\ 0 & \lambda \end{bmatrix}$$

$$A^T A - \lambda I = \begin{bmatrix} 2-\lambda & 0 \\ 0 & 3-\lambda \end{bmatrix}$$

$$|A^T A - \lambda I| = 0$$

$$(2-\lambda)(3-\lambda) = 0$$

$$\lambda_1 = 3, \lambda_2 = 2$$

for $\lambda_1 = 3$

$$A^T A - \lambda_1 I = A^T A - 3I = \begin{bmatrix} -1 & 0 \\ 0 & 0 \end{bmatrix}$$

$$[A^T A - 3I] \vec{v}_1 = 0$$

$$\begin{bmatrix} -1 & 0 \\ 0 & 0 \end{bmatrix} \begin{bmatrix} v_{11} \\ v_{12} \end{bmatrix} = 0$$

Let $v_{12} = t$ then $v_{11} = 0$

$$\text{Thus, } \vec{v}_1 = \begin{bmatrix} v_{11} \\ v_{12} \end{bmatrix} = \begin{bmatrix} 0 \\ 1 \end{bmatrix}$$

for $\lambda_2 = 2$

$$A^T A - \lambda_2 I = A^T A - 2I = \begin{bmatrix} 0 & 0 \\ 0 & 1 \end{bmatrix}$$

$$[A^T A - 2I] \vec{v}_2 = 0$$

$$\begin{bmatrix} 0 & 0 \\ 0 & 1 \end{bmatrix} \begin{bmatrix} v_{21} \\ v_{22} \end{bmatrix} = 0$$

Let $v_{21} = -1$

then $v_{22} = 0$

$$\vec{v}_2 = \begin{bmatrix} v_{21} \\ v_{22} \end{bmatrix} = \begin{bmatrix} -1 \\ 0 \end{bmatrix}$$

Normalized vectors are

$$\vec{v}_1 = \begin{bmatrix} 0 \\ 1 \end{bmatrix}, \quad \vec{v}_2 = \begin{bmatrix} -1 \\ 0 \end{bmatrix}$$

$$V = \begin{bmatrix} 0 & -1 \\ 1 & 0 \end{bmatrix} \quad V^T = \begin{bmatrix} 0 & 1 \\ -1 & 0 \end{bmatrix}$$

$$\Sigma = \begin{bmatrix} \sigma_1 & 0 \\ 0 & \sigma_2 \\ 0 & 0 \end{bmatrix} = \begin{bmatrix} \sqrt{3} & 0 \\ 0 & \sqrt{2} \\ 0 & 0 \end{bmatrix}$$

$$A A^T - \lambda I = \begin{bmatrix} 2-\lambda & 1 & 0 \\ 0 & 1-\lambda & 1 \\ 0 & 1 & 2-\lambda \end{bmatrix}$$

$$[A A^T - \lambda I] = 0$$

$$\lambda^3 - S_1 \lambda^2 + S_2 \lambda - S_3 = 0$$

$$S_1 = 2 + 1 + 2 = 5$$

$$S_2 = (2-1) + (4) + (2-1) = 6$$

$$S_3 = 0$$

$$\begin{aligned} |A A^T - \lambda I| &= \lambda^3 - 5\lambda^2 + 6\lambda = 0 \\ &\lambda (\lambda^2 - 5\lambda + 6) = 0 \\ &\lambda (\lambda^2 - 3\lambda - 2\lambda + 6) = 0 \\ &\lambda (\lambda (\lambda - 3) - 2(\lambda - 3)) = 0 \\ &\lambda (\lambda - 3) (\lambda - 2) = 0 \end{aligned}$$

Hence Eigen Values are,

$$\lambda_1 = 3, \quad \lambda_2 = 2, \quad \lambda_3 = 0$$

for $\lambda_1 = 3$

$$A A^T - \lambda_1 I = A A^T - 3I = \begin{bmatrix} -1 & 1 & 0 \\ 1 & -2 & 1 \\ 0 & 1 & -1 \end{bmatrix}$$

$$(A A^T - 3I) \vec{u}_1 = 0$$

$$\begin{bmatrix} -1 & 1 & 0 \\ 1 & -2 & 1 \\ 0 & 1 & -1 \end{bmatrix} \begin{bmatrix} u_{11} \\ u_{12} \\ u_{13} \end{bmatrix} = 0$$

$$-u_{11} + u_{12} + u_{13} = 0$$

$$u_{11} - 2u_{12} + u_{13} = 0$$

$$\frac{u_{11}}{\begin{pmatrix} 1 & 0 \\ -2 & 1 \end{pmatrix}} = \frac{-u_{12}}{\begin{pmatrix} -1 & 0 \\ 1 & 1 \end{pmatrix}} = \frac{u_{13}}{\begin{pmatrix} -1 & 1 \\ 1 & -2 \end{pmatrix}}$$

$$\frac{u_{11}}{1} = \frac{-u_{12}}{-1} = \frac{u_{13}}{1}$$

$$\frac{u_{11}}{1} = \frac{u_{12}}{1} = \frac{u_{13}}{1}$$

$$\vec{u}_1 = \begin{bmatrix} u_{11} \\ u_{12} \\ u_{13} \end{bmatrix} = \begin{bmatrix} 1 \\ 1 \\ 1 \end{bmatrix} = \begin{bmatrix} 1/\sqrt{3} \\ 1/\sqrt{3} \\ 1/\sqrt{3} \end{bmatrix}$$

for $\lambda_2 = 2$,

$$A A^T - \lambda_2 I = A A^T - 2I = 0$$

$$A A^T - 2I = \begin{bmatrix} 0 & 1 & 0 \\ 1 & -1 & 1 \\ 0 & 1 & 0 \end{bmatrix}$$

$$[A A^T - 2I] \vec{u}_2 = 0$$

$$\begin{bmatrix} 0 & 1 & 0 \\ 1 & -1 & 1 \\ 0 & 1 & 0 \end{bmatrix} \begin{bmatrix} u_{21} \\ u_{22} \\ u_{23} \end{bmatrix} = 0$$

$$0 \cdot u_{21} + u_{22} + 0 \cdot u_{23} = 0$$

$$u_{21} - u_{22} + u_{23} = 0$$

$$\frac{u_{21}}{\begin{pmatrix} 1 & 0 \\ -1 & 1 \end{pmatrix}} = \frac{-u_{22}}{\begin{pmatrix} 0 & 0 \\ 1 & 1 \end{pmatrix}} = \frac{u_{23}}{\begin{pmatrix} 0 & 1 \\ 1 & -1 \end{pmatrix}}$$

$$\frac{u_{21}}{-1} = \frac{-u_{22}}{0} = \frac{u_{23}}{+1}$$

$$\frac{u_{21}}{-1} = \frac{u_{22}}{0} = \frac{u_{23}}{+1}$$

$$\vec{u}_2 = \begin{bmatrix} u_{21} \\ u_{22} \\ u_{23} \end{bmatrix} = \begin{bmatrix} -1 \\ 0 \\ +1 \end{bmatrix} = \begin{bmatrix} -1/\sqrt{2} \\ 0 \\ +1/\sqrt{2} \end{bmatrix}$$

~~2~~

for $\lambda_3 = 0$,

$$A A^T - \lambda_3 I = A A^T - 0I = 0$$

$$A A^T - 0I = \begin{bmatrix} 2 & 1 & 0 \\ 1 & 1 & 1 \\ 0 & 1 & 2 \end{bmatrix}$$

$$(A A^T - 0I) \vec{u}_3 = 0$$

$$\begin{bmatrix} 2 & 1 & 0 \\ 1 & 1 & 1 \\ 0 & 1 & 2 \end{bmatrix} \begin{bmatrix} u_{31} \\ u_{32} \\ u_{33} \end{bmatrix} = 0$$

$$2u_{31} + u_{32} + 0 \cdot u_{33} = 0$$

$$u_{31} + u_{32} + u_{33} = 0$$

$$\frac{u_{31}}{\begin{vmatrix} 1 & 0 \\ 1 & 1 \end{vmatrix}} = \frac{-u_{32}}{\begin{vmatrix} 2 & 0 \\ 1 & 1 \end{vmatrix}} = \frac{u_{33}}{\begin{vmatrix} 2 & 1 \\ 1 & 1 \end{vmatrix}}$$

$$\frac{u_{31}}{1} = \frac{+u_{32}}{-2} = \frac{u_{33}}{1}$$

$$\vec{u}_3 = \begin{bmatrix} u_{31} \\ u_{32} \\ u_{33} \end{bmatrix} = \begin{bmatrix} 1 \\ -2 \\ 1 \end{bmatrix} = \begin{bmatrix} \frac{1}{\sqrt{6}} \\ -\frac{2}{\sqrt{6}} \\ \frac{1}{\sqrt{6}} \end{bmatrix}$$

$$U = \begin{bmatrix} \frac{1}{\sqrt{3}} & -\frac{1}{\sqrt{2}} & \frac{1}{\sqrt{6}} \\ \frac{1}{\sqrt{3}} & 0 & -\frac{2}{\sqrt{6}} \\ \frac{1}{\sqrt{3}} & +\frac{1}{\sqrt{2}} & \frac{1}{\sqrt{6}} \end{bmatrix}$$

$$U \epsilon V^T = \begin{bmatrix} \frac{1}{\sqrt{3}} & -\frac{1}{\sqrt{2}} & \frac{1}{\sqrt{6}} \\ \frac{1}{\sqrt{3}} & 0 & -\frac{2}{\sqrt{6}} \\ \frac{1}{\sqrt{3}} & \frac{1}{\sqrt{2}} & \frac{1}{\sqrt{6}} \end{bmatrix} \begin{bmatrix} \sqrt{3} & 0 \\ 0 & \sqrt{2} \\ 0 & 0 \end{bmatrix} \begin{bmatrix} 0 & 1 \\ -1 & 0 \end{bmatrix}$$



```
# Singular Value Decomposition
from numpy import array
from scipy.linalg import svd
```

```
# define a matrix
```

```
A = array([[3, 1, 1], [-1, 3, 1]])
print("A =", A)
```

```
# SVD
```

```
u, s, v_T = svd(A)
```

```
# left singular vectors
```

```
print("u = ")
print(u)
```

```
# singular values
```

```
print("s = ")
print(s)
```

```
# right singular vectors,
```

```
print("v_T = ")
print(v_T)
```

4b.

$$A = \begin{bmatrix} -2 & 2 & -3 \\ 2 & 1 & -6 \\ -1 & -2 & 0 \end{bmatrix}$$

$$A - \lambda I = \begin{bmatrix} -2-\lambda & 2 & -3 \\ 2 & 1-\lambda & -6 \\ -1 & -2 & -\lambda \end{bmatrix}$$

$$|A - \lambda I| = 0$$

$$\lambda^3 - s_1 \lambda^2 + s_2 \lambda - s_3 = 0$$

$$s_1 = -2 + 1 = -1$$

$$s_2 = (-12) - (3) + (-2 - 4) = -21$$

$$s_3 = 45$$

$$\lambda^3 + \lambda^2 - 21\lambda - 45 = 0$$

$$5 \left| \begin{array}{cccc|c} 1 & 1 & -21 & -45 & \\ 0 & 5 & 30 & 45 & \\ \hline 1 & 6 & 9 & 0 & \end{array} \right.$$

$$\lambda^2 - 6\lambda + 9 = 0$$

$$\lambda^2 - 3\lambda - 3\lambda + 9 = 0$$

$$\lambda(\lambda - 3) - 3(\lambda - 3) = 0$$

$$(\lambda - 3)(\lambda - 3) = 0$$

$$\lambda_1 = 5, \quad \lambda_2 = 3, \quad \lambda_3 = 3$$

for $\lambda_1 = 5$,

$$A - 5I = \begin{bmatrix} -7 & 2 & -3 \\ 2 & -4 & -6 \\ -1 & -2 & -5 \end{bmatrix}$$

$$[A - 5I] \vec{x} = 0$$

$$[A - 5I] \begin{bmatrix} x_1 \\ x_2 \\ x_3 \end{bmatrix} = 0$$

$$-7x_1 + 2x_2 - 3x_3 = 0$$

$$2x_1 - 4x_2 - 6x_3 = 0$$

$$\frac{x_1}{\begin{vmatrix} 2 & -3 \\ -4 & -6 \end{vmatrix}} = \frac{-x_2}{\begin{vmatrix} -7 & -3 \\ 2 & -6 \end{vmatrix}} = \frac{x_3}{\begin{vmatrix} -7 & 2 \\ 2 & -4 \end{vmatrix}}$$

$$\frac{x_1}{-12-12} = \frac{-x_2}{42+6} = \frac{x_3}{28-4}$$

$$\frac{x_1}{-24} = \frac{x_2}{-48} = \frac{x_3}{24}$$

$$\vec{x} = \begin{bmatrix} -24 \\ -48 \\ 24 \end{bmatrix} = \begin{bmatrix} -1 \\ -2 \\ 1 \end{bmatrix}$$

$$[A + 3I] \vec{y} = 0$$

$$\begin{bmatrix} 1 & 2 & -3 \\ 2 & 4 & -6 \\ -1 & -2 & 3 \end{bmatrix} \begin{bmatrix} y_1 \\ y_2 \\ y_3 \end{bmatrix} = 0$$

Reduced row Echelon form of matrix is,

$$\begin{bmatrix} 1 & 2 & -3 \\ 0 & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix} \begin{bmatrix} y_1 \\ y_2 \\ y_3 \end{bmatrix} = 0$$

Let $y_2 = t$, $y_3 = s$ then $y_1 = 3s - 2t$

$$\vec{y} = \begin{bmatrix} 3s - 2t \\ t \\ s \end{bmatrix} = \begin{bmatrix} -2 \\ 1 \\ 0 \end{bmatrix} t + \begin{bmatrix} 3 \\ 0 \\ 1 \end{bmatrix} s$$

$$\vec{y} = \begin{bmatrix} -2 \\ 1 \\ 0 \end{bmatrix} \begin{bmatrix} 3 \\ 0 \\ 1 \end{bmatrix}$$

$$P = \begin{bmatrix} -1 & -2 & 3 \\ -2 & 1 & 0 \\ 1 & 0 & 1 \end{bmatrix}$$

2

$$P^{-1} = \begin{bmatrix} -1/8 & -1/4 & 3/8 \\ -1/4 & 1/2 & 3/4 \\ 1/8 & 1/4 & 5/8 \end{bmatrix}$$

$$D = \begin{bmatrix} 5 & 0 & 0 \\ 0 & -3 & 0 \\ 0 & 0 & -3 \end{bmatrix}$$

$$A = P D P^{-1}$$

$$\begin{bmatrix} -2 & 2 & -3 \\ 2 & 1 & -6 \\ -1 & -2 & 0 \end{bmatrix} = \begin{bmatrix} -1 & -2 & 3 \\ -2 & 1 & 0 \\ 1 & 0 & 1 \end{bmatrix} \begin{bmatrix} 5 & 0 & 0 \\ 0 & -3 & 0 \\ 0 & 0 & -3 \end{bmatrix} \begin{bmatrix} -1/8 & -1/4 & 3/8 \\ -1/4 & 1/2 & 3/4 \\ 1/8 & 1/4 & 5/8 \end{bmatrix}$$

5 a. Signal: A function of one or more independent variables which contains some information is called signal.

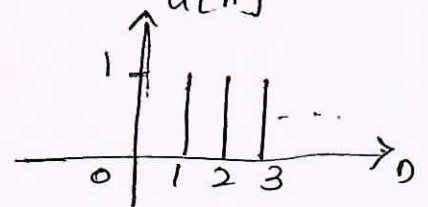
System: A system is a set of elements or functional blocks that are connected together and produces an output in response to an input signal.

Discrete elementary signals are,

i) Unit Step function:

The unit step signal has amplitude of "one" for positive value and "zero" amplitude for negative values.

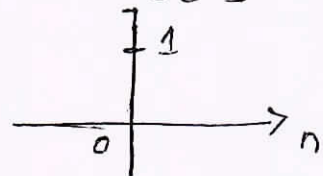
$$u(n) = \begin{cases} 1 & \text{for } n \geq 0 \\ 0 & \text{for } n < 0 \end{cases}$$



(ii) Unit sample signal $\delta(n)$

Amplitude of unit sample is 1, at $n=0$ & it has zero values at all other values of n .

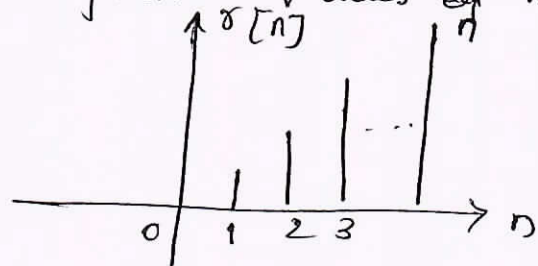
$$\delta[n] = \begin{cases} 1 & \text{for } n=0 \\ 0 & \text{for } n \neq 0 \end{cases}$$



(iii) Unit Ramp function

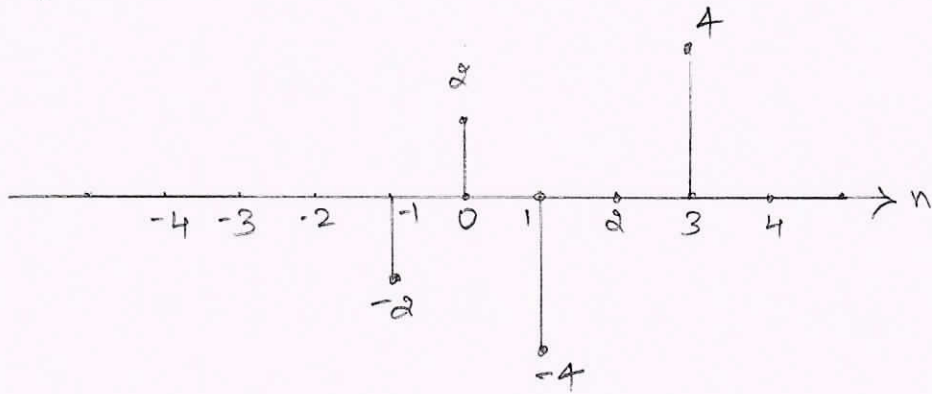
The amplitude of every sample increases linearly with its number (n) for positive values of n .

$$r(n) = \begin{cases} n & \text{for } n \geq 0 \\ 0 & \text{for } n < 0 \end{cases}$$

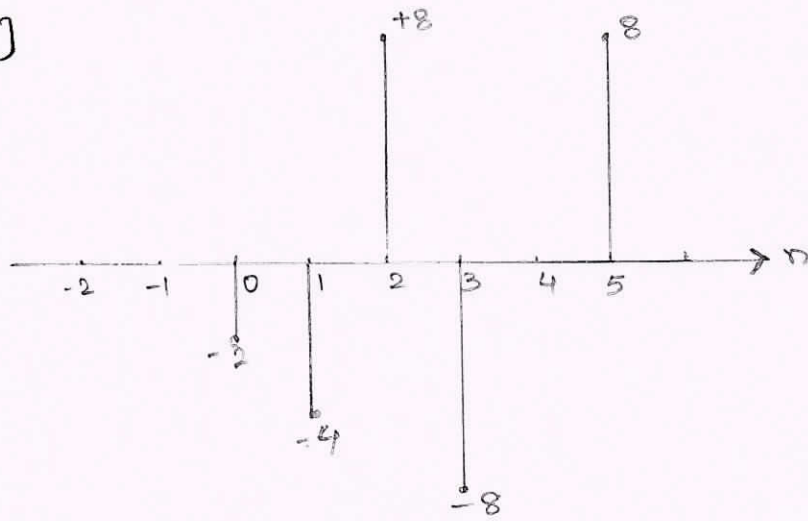


5b

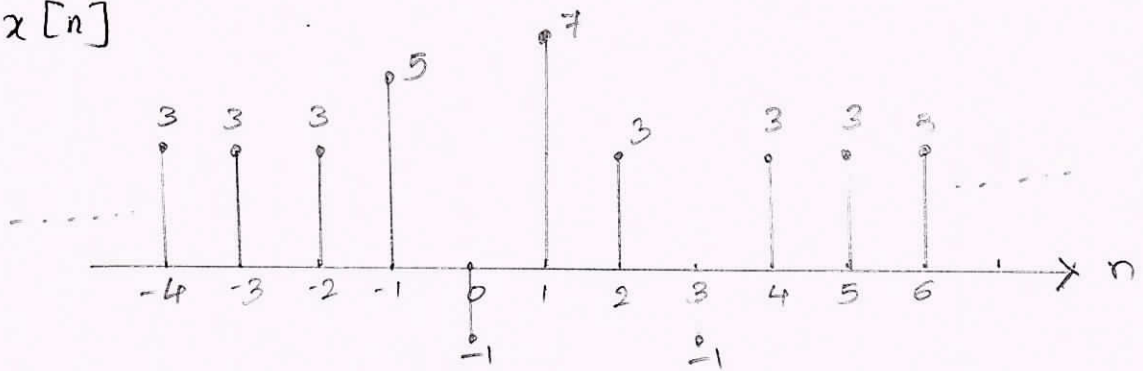
$x[n]$



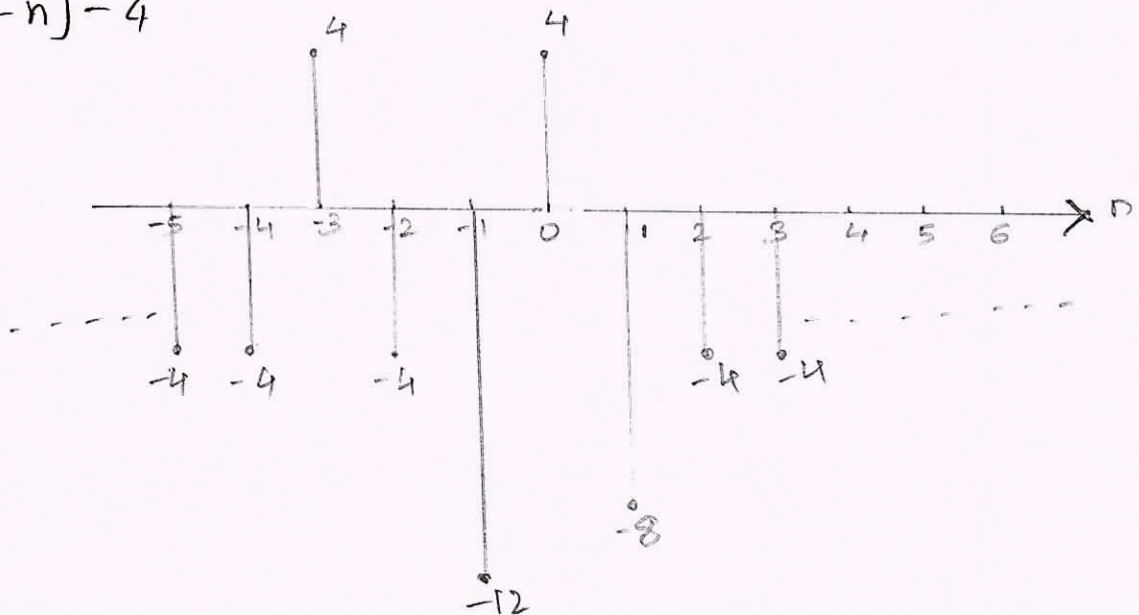
$2x[n-2]$



$3-x[n]$



$2x[-n]-4$



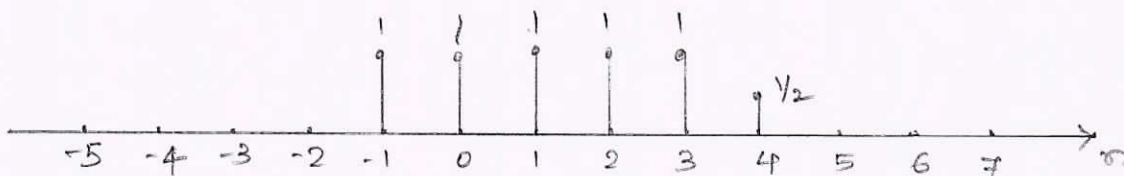
[Handwritten signature]

5c. Sketch $x[n] = \begin{cases} 1 & ; & -1 \leq n \leq 3 \\ \frac{1}{2} & ; & n=4 \\ 0 & ; & \text{otherwise} \end{cases}$

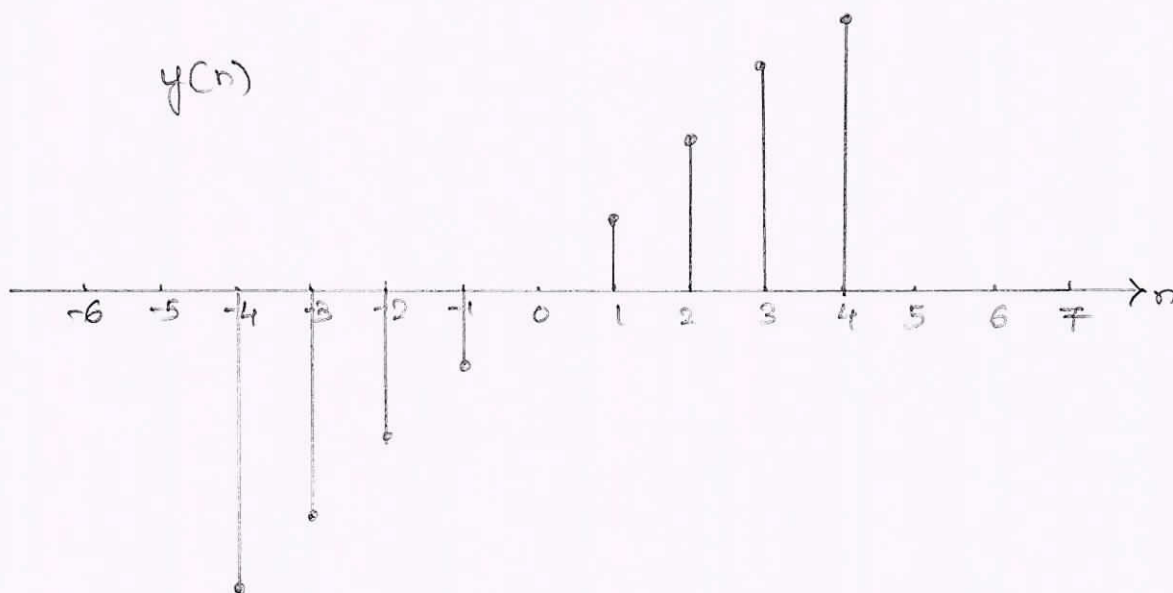
$y[n] = \begin{cases} \frac{1}{2}n & ; & |n| \leq 4 \\ 0 & ; & \text{otherwise} \end{cases}$

Also sketch $x[n+2]$ $y[1-2n]$

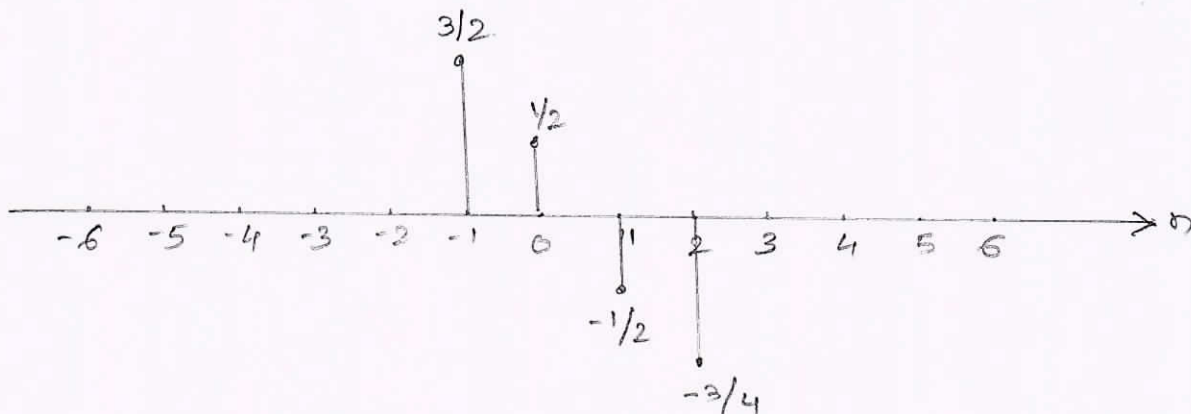
$x(n)$



$y(n)$



$x(n+2)$ $y(1-2n)$



[Handwritten mark]

6 a. (i) $y[n] = 2x[n] + \frac{1}{x[n-2]}$

The $y[n]$ depends upon previous and present input. Hence it is causal system.

$$y[n] = T \left\{ 2x[n] + \frac{1}{x[n-2]} \right\}$$

$$y_3[n] \neq T y_3'[n]$$

\therefore System is non linear

Since $y[n]$ is a bounded output because $x[n]$ is also bounded therefore system is stable.

Let us apply the delayed input to the system.

$$y[n, k] = T \left\{ 2(n-k) + \frac{1}{n-k-2} \right\}$$

$$y[n, k] = T \left\{ 2[n-k] + \frac{1}{n-k-2} \right\}$$

\therefore hence $y(n, k) = y(n-k)$

there system is time invariant

• system is memory system

ii) $y[n] = \ln [3 + |x[n]|]$

$$y_1[n] = \ln [3 + |x_1[n]|]$$

$$y_2[n] = \ln [3 + |x_2[n]|]$$

$$y_3[n] = T [a_1 \ln [3 + |x_1[n]|] + a_2 \ln [3 + |x_2[n]|]]$$

$$y_3'[n] = a_1 y_1[n] + a_2 y_2[n]$$

$$= a_1 \ln [3 + |x_1[n]|] + a_2 \ln [3 + |x_2[n]|]$$

$$y_3[n] = \ln [a_1 [3 + |x_1[n]|] + a_2 [3 + |x_2[n]|]]$$

$$y_3'[n] \neq y_3[n]$$

\therefore system is non linear.

Since $y[n]$ not depends upon the present input
Hence system is noncausal.

$$y[n, k] = T \{ \ln [3 + |a[n-k]|] \}$$

$$y[n, k] = T \{ \ln [3 + |a'(n-k)|] \}$$

$$\text{hence } y[n-k] = y[n-k]$$

The system is time invariant

Since $x[n]$ is a bounded input, hence $y[n]$ is a bounded output. Hence system is stable
It is memoryless system.

(iii) $y[n] = \cos x[n]$

Since $y[n]$ depends upon the present input, Hence the system is causal.

$$y_1[n] = \cos x_1[n]$$

$$y_2[n] = \cos x_2[n]$$

$$y_3[n] = \cos [x_1[n] + x_2[n]]$$

$$y_3'[n] = \cos x_1[n] + \cos x_2[n]$$

$$y_3[n] \neq y_3'[n]$$

The system is non linear
Since $x[n]$ is bounded input have $y[n]$ is bounded output.

Hence system is stable.

$$y[n, k] = \cos x[n-k]$$

$$y[n-k] = \cos x[n-k]$$

$$y[n-k] = y[n-k]$$

Hence system is time invariant.

It is memoryless system.

Q.E.D.

iv) $y[n] = \gamma^n x[n] \quad ; \quad \gamma > 1$

$y[n]$ depends upon the present input. Hence system is causal.

Here $\gamma > 1$, Hence $n \rightarrow \infty$, $\gamma^n \rightarrow \infty$

Therefore $y[n]$ will be unbounded even if $x[n]$ is bounded. Hence this is unstable.

$$y[n] = T \{ \gamma^n x[n] \}$$

$$y_1[n] = T \{ x_1[n] \} = \gamma^n x_1[n]$$

$$y_2[n] = T \{ x_2[n] \} = \gamma^n x_2[n]$$

$$y_3[n] = T \{ a_1 x_1[n] + a_2 x_2[n] \} \\ = a_1 \gamma^n x_1[n] + a_2 \gamma^n x_2[n]$$

$$y_3'[n] = a_1 y_1[n] + a_2 y_2[n]$$

$$y_3'[n] = a_1 \gamma^n x_1[n] + a_2 \gamma^n x_2[n]$$

$$\therefore y_3[n] = y_3'[n]$$

\therefore The system is linear

$$y[n, k] = T \{ x[n-k] \}$$

$$= \gamma^n x[n-k]$$

$$y[n-k] = \gamma^{n-k} x[n-k]$$

$$y[n, k] \neq y[n-k]$$

Hence the system is time variant
It is memory less system.

~~Q~~

```

6 b. import numpy as np
import matplotlib.pyplot as plt

def exponential(a, n):
    expo = [n]
    for sample in n:
        expo.append(np.exp(a * sample))
    return (expo)

a = 2
UL = 1
LL = -1
n = np.arange(LL, UL, 0.1)
x = exponential(a, n)
plt.stem(n, x)
plt.xlabel('n')
plt.xticks(np.arange(LL, UL, 0.2))
# plt.yticks([0, UL, 1])
plt.ylabel('x[n]')
plt.title('Exponential signal  $e^{ax}(an)$ ')
plt.savefig("Exponential.png")

```


6 b

Triangular waveform Generation

import matplotlib.pyplot as plt

import numpy as np

% matplotlib inline

x = np.arange(0, 101)

def tri(x):

res = x%.50

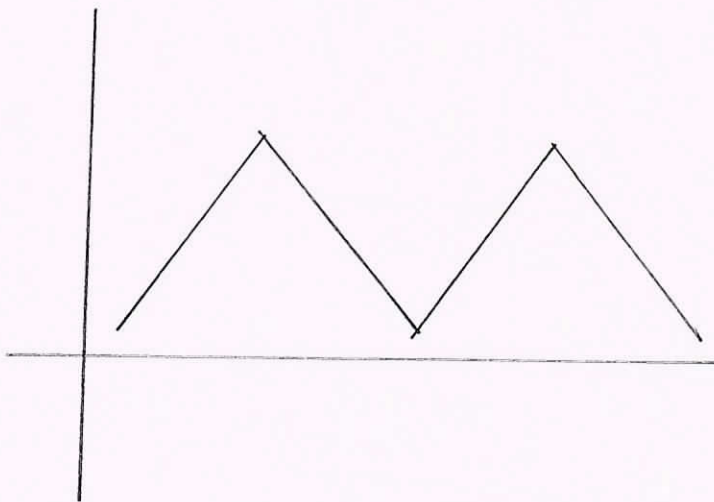
if res < 25:

return res

else:

return 50 - res

plt.plot(x, list(map(tri, x)))



2/

Module 4

$$y[n] = \sum_{k=-\infty}^{\infty} x[n] h[n-k]$$

$$= \sum_{k=-\infty}^{\infty} a^n u[k] b^{n-k} u[n-k]$$

$$u[k] = \begin{cases} 1 & \text{for } k \geq 0 \\ 0 & \text{for } k < 0 \end{cases}$$

$$y[n] = \sum_{k=0}^n a^k b^{n-k} u[n-k]$$

$$u[n-k] = \begin{cases} 1 & \text{for } n \geq k \\ 0 & \text{for } n < k \text{ or } k > n \end{cases}$$

$$y[n] = \sum_{k=0}^n a^k b^{n-k} \text{ for } n \geq k \geq 0$$

$$= \sum_{k=0}^n a^k \cdot b^k \cdot b^{-k}$$

$$= b^n \sum_{k=0}^n (a \cdot b^{-1})^k$$

$$y[n] = b^n \frac{(ab^{-1})^{n+1} - 1}{(ab^{-1}) - 1}$$

$$\therefore \sum_{k=0}^n r^k = \frac{r^{n+1} - 1}{r - 1}$$

$$y[n] = b^n \frac{(a/b)^{n+1} - 1}{a/b - 1}$$

$$= b^n \cdot b \frac{(a/b)^{n+1} - 1}{a - b}$$

$$= \frac{b^{n+1} \left(\frac{a^{n+1}}{b^{n+1}} - 1 \right)}{a - b}$$

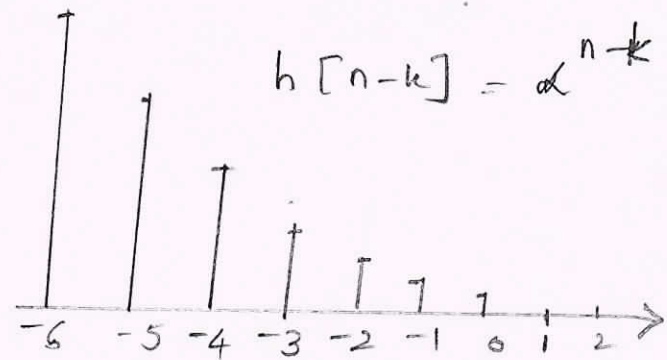
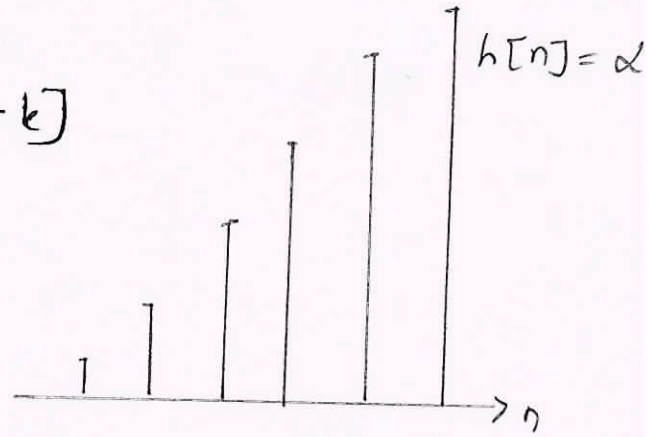
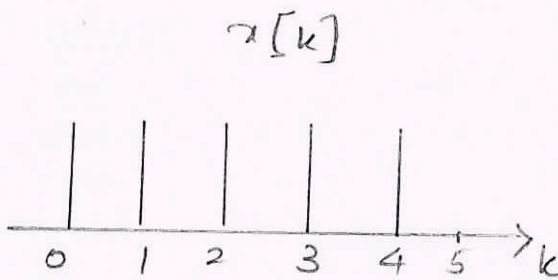
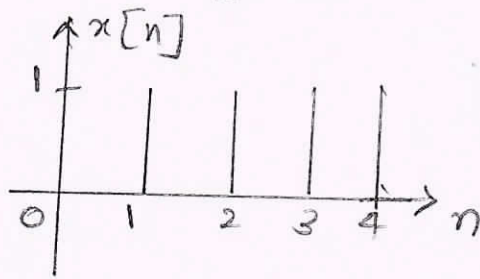
$$y[n] = \frac{a^{n+1} - b^{n+1}}{a - b} \text{ for } n \geq 0 \text{ and } a \neq b.$$

Q //

b

$$y[n] = x[n] * h[n]$$

$$y[n] = \sum_{k=-\infty}^{\infty} x[k] h[n-k]$$



Case 1)

$$n < 0$$

$$y[n] = 0;$$

Case 2)

$$n \geq 0, n \leq 4 \quad y[n] = \frac{1 - \alpha^{n+1}}{1 - \alpha}$$

Case 3)

$$n > 4 \quad n - 6 \leq 0 \quad (4 < n \leq 6)$$

$$y[n] = \frac{\alpha^{n-4} - \alpha^{n+1}}{1 - \alpha}$$

Case 4)

$$n > 6 \quad n - 6 < 4 \quad (6 < n \leq 10)$$

$$y[n] = \frac{\alpha^{n-4} - \alpha^7}{1 - \alpha}$$

Case 5)

$$n - 6 > 4 \quad (n > 10) \quad y[n] = 0$$

Q.E.D.

$$y[n] = \begin{cases} 0 & ; n < 0 \\ 1 - \frac{\alpha^{n+1}}{1-\alpha} & ; 0 \leq n \leq 4 \\ \alpha^{n-4} - \frac{\alpha^{n+1}}{1-\alpha} & ; 4 < n \leq 6 \end{cases}$$

$$y[n] = \begin{cases} \alpha^{n-4} - \frac{\alpha^7}{1-\alpha} & ; 6 < n \leq 10 \\ 0 & ; n > 10 \end{cases}$$

a)
@ (*) $h(n) = e^{-n} \cos(n) \cdot u(n)$

Multiplication of $u[n]$ ensures that $h[n] = 0$ for $n < 0$
Hence given system is causal.

$$\sum_{n=-\infty}^{\infty} |h[n]| = \sum_{n=-\infty}^{\infty} e^{-n} \cos(n) \cdot u(n)$$

$$= \sum_{n=0}^{\infty} |\cos[n]| \cdot e^{-n}$$

$$-1 \leq \cos[n] \leq 1$$

$$|\cos[n]| < 1$$

$$\sum_{n=-\infty}^{\infty} |h[n]| = \sum_{n=0}^{\infty} |\cos[n]| \cdot e^{-n}$$

for every n , value of $\cos[n]$ is finite, but summation is going for $n \rightarrow \infty$, which causes

$$\sum_{n=-\infty}^{\infty} |h[n]| \rightarrow \infty$$

So, given system is not stable

$$(b) (i) \quad h[n] = (0.99)^n u(n+3)$$

- System is dynamic, It requires memory since $h[n]$ is not in terms of $\delta(n)$
- The system is noncausal since $h(n) \neq 0$ for $n < 0$ because of presence of $u(n+3)$

• Consider $\sum_{k=-\infty}^{\infty} |h(k)| = \sum_{k=-\infty}^{\infty} (0.99)^k u(k+3)$

$$= \sum_{k=-3}^{\infty} (0.99)^k \quad \text{since } u(k+3) = \begin{cases} 1 & \text{for } k \geq -3 \\ 0 & \text{for } k < -3 \end{cases}$$

$$= (0.99)^{-3} + (0.99)^{-2} + (0.99)^{-1} + \sum_{k=0}^{\infty} (0.99)^k$$

$$= 3.061 + \frac{1}{1-0.99} \quad \text{By geometric series}$$

$$= 103.061 \quad \text{which is finite.}$$

Since impulse response is absolutely summable the system is stable.

8 (a)
(c)

$$h[n] = n \left(\frac{1}{2}\right)^n u[n]$$

(i) To check stability

The discrete time system is stable if

$$\sum_{k=-\infty}^{\infty} |h[k]| < \infty$$

Let us check this condition for given system, i.e.

$$\sum_{k=-\infty}^{\infty} |h[k]| = \sum_{k=-\infty}^{\infty} k \left[\frac{1}{2}\right]^k u[k]$$

Since $u[k] = 1$ for $k \geq 0$, above equation

becomes,

$$\sum_{k=-\infty}^{\infty} |h[k]| = \sum_{k=0}^{\infty} k \left(\frac{1}{2}\right)^k$$

$$\sum_{k=0}^{\infty} k \cdot a^k = \frac{a}{(1-a)^2}, \text{ if } |a| < 1$$

Then, above summation becomes,

$$\sum_{k=-\infty}^{\infty} |h[k]| = \frac{\frac{1}{2}}{\left(1-\frac{1}{2}\right)^2} = 2$$

$$\text{Then } \sum_{k=-\infty}^{\infty} |h[k]| = 2 < \infty$$

hence the system is stable.

(ii) To check for Causality

A LTI system is causal if
 $h[n] = 0$ for $n < 0$

given impulse response is
 $h[n] = n \left(\frac{1}{2}\right)^n u[n]$

✗

$$u[n] = 1 \text{ for } n \geq 0$$

$$u[n] = 0 \text{ for } n < 0$$

Equation will be

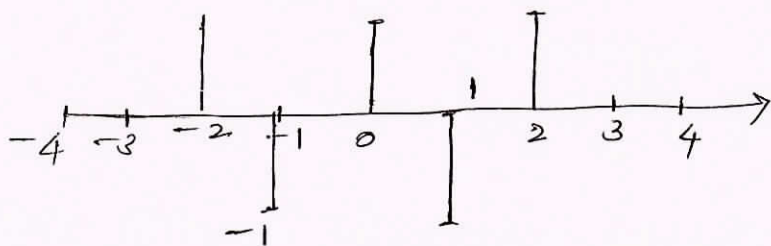
$$h[n] = \begin{cases} n \left(\frac{1}{2}\right)^n & \text{for } n \geq 0 \\ 0 & \text{for } n < 0 \end{cases}$$

Thus the system is causal since $h[n] = 0, n < 0$

8 b.

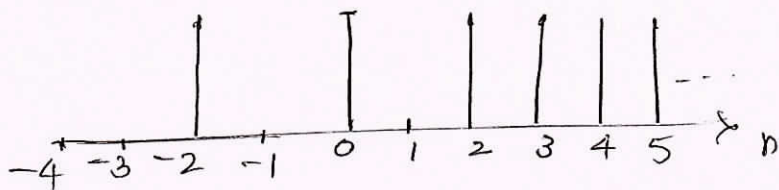
$$h[n] = (-1)^n \{ u(n+2) - u(n-3) \}$$

$h[n]$



$$s[n] = \begin{cases} 0 & ; \quad n \leq -3 \\ 1 & ; \quad n = -2, 0, 2 \\ 0 & ; \quad n = -1, 1 \end{cases}$$

plot of $s[n]$



```
# Step Response from given impulse response.  
import numpy as np.  
import matplotlib.pyplot as plt
```

```
# Given impulse response  
impulse_response = [0.1, 0.2, 0.3, 0.4]
```

```
# convert the impulse response to a step response  
step_response = np.cumsum(impulse_response)
```

```
# Plot the step response
```

```
plt.plot(step_response)
```

```
plt.xlabel('Time')
```

```
plt.ylabel('Response')
```

```
plt.title('Step Response from Impulse Response')
```

```
plt.grid()
```

```
plt.show()
```

9 a Z transform is a mathematical tool, which is used to convert difference equations in time domain, into algebraic equations in Z domain.

Z transform of $x(n)$ is denoted by $X(z)$ which is defined as,

$$X(z) = \sum_{n=-\infty}^{\infty} x(n) z^{-n} \quad \text{--- (A)}$$

Types of Z transform

- (i) Bilateral Z transform: The Z transform defined in eqⁿ (A) has both sided summation, hence it is called bilateral Z transform,
- (ii) Unilateral Z transform: It is defined as,

$$X(z) = \sum_{n=0}^{\infty} x(n) z^{-n}$$

Here summation is from 0 to ∞ . That is one sided.

Properties of ROC

- ROC cannot contain any poles.
- If $x[n]$ is finite causal sequence then ROC is entire Z plane except at $z=0$
- If $x[n]$ is finite non causal sequence, then ROC is entire Z plane except at $z=\infty$
- If $x[n]$ is finite double sided sequence, then ROC is entire Z plane except at $z=0$ and ∞ .
- If $x[n]$ is causal infinite length sequence then ROC is of the form, $|z| > r_{\max}$, where r_{\max} equals largest magnitude of any of the poles of $X(z)$

→ If $x[n]$ is non causal infinite length sequence then ROC is of the form $|z| < r_{min}$,

→ If $x[n]$ is two sided sequence of infinite duration, then ROC is of the form $r_1 < |z| < r_2$

→ ROC of an LTI stable system contains the unit circle, in the z plane.

→ ROC must be connected region (circle)

q b. $x[n] = \left(\frac{1}{2}\right)^n$

(i) when $n=0$, $x[n]=1$

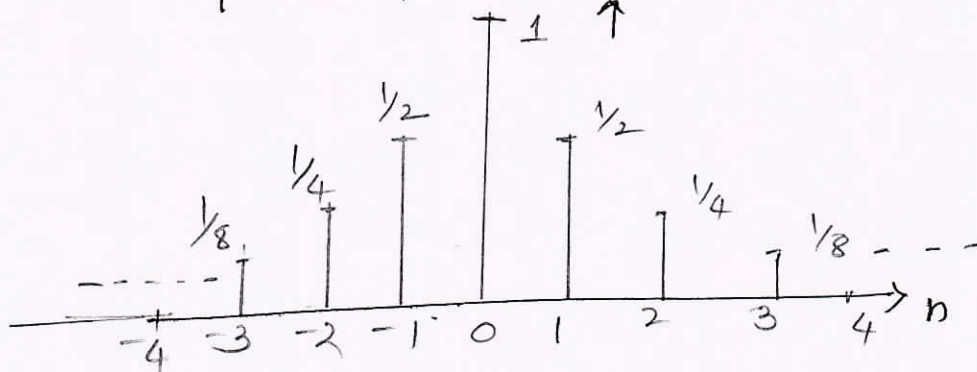
$n=1$ $x[n]=\frac{1}{2}$

$n=2$ $x[n]=\frac{1}{4}$

$n=-1$ $x[n]=\frac{1}{2}$

$n=-2$ $x[n]=\frac{1}{4}$

$x[n] = \{ \dots, \frac{1}{4}, \frac{1}{2}, 1, \frac{1}{2}, \frac{1}{4}, \dots \}$



(ii) $x(n) = \frac{1}{2}^{(n)}$

$x(n) = \left(\frac{1}{2}\right)^n u(n) + \left(\frac{1}{2}\right)^{-n} u(-n-1)$

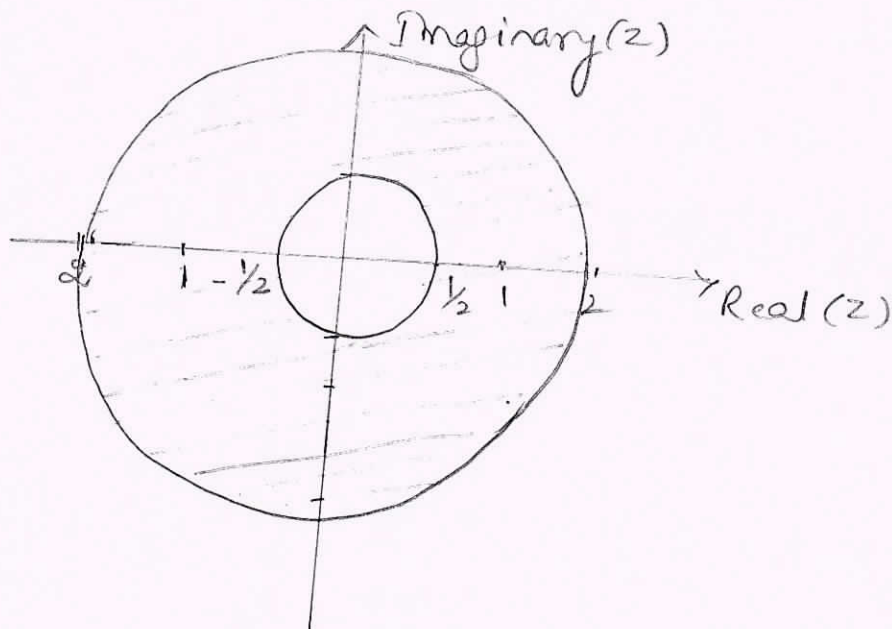
$X(z) = \frac{z}{z - \frac{1}{2}} - \frac{z}{z - 2}$

ROC = $\frac{1}{2} < |z| < 2$

$$\begin{aligned}
 &= \frac{z(z-2) - z(z-\frac{1}{2})}{(z-\frac{1}{2})(z-2)} \\
 &= \frac{z^2 - 2z - z^2 + \frac{1}{2}z}{z^2 - 2z - \frac{1}{2}z + 1} \\
 &= \frac{-3z/2}{z^2 - 5/2z + 1}
 \end{aligned}$$

$$X(z) = \frac{-3z}{2(z-\frac{1}{2})(z-2)}$$

$$\text{ROC} = \frac{1}{2} < |z| < 2$$



c. $x(n) = (\frac{1}{2})^n u(n) * (\frac{1}{3})^n u(n)$

$$X(z) = \frac{1}{1 - (\frac{1}{2})z^{-1}} \cdot \frac{1}{1 - (\frac{1}{3})z^{-1}}$$

$$X(z) = \frac{1}{(1 - \frac{1}{2}z^{-1})(1 - \frac{1}{3}z^{-1})}$$

$$\text{ROC} : |z| > \frac{1}{2}$$

$$|z| > \frac{1}{3}$$

10 a) Convolution:

(i) If $Z \{ x[n] \}$; $R_x^- < |z| < R_x^+$

and $Z \{ y[n] \}$; $R_y^- < |z| < R_y^+$

$$\text{Then } Z \{ x[n] y[n] \} = \frac{1}{2\pi j} \oint_{C_1} x(v) y(z/v) v^{-1} dv$$

where \oint_{C_1} is complex contour integral and C_1 is closed contour in the intersection of the ROC's of $x(v)$ and $1/(z/v)$

ROC for resulting transform is,

$$R_x^- R_y^- < |z| < R_x^+ R_y^+$$

$$Z \{ x[n] y[n] \} = \sum_{n=-\infty}^{\infty} x[n] y[n] z^{-n}$$

$$x[n] = \frac{1}{2\pi j} \oint_{C_1} x[z] \frac{z^n}{z} dz$$

$$Z \{ x[n] y[n] \} = \sum_{n=-\infty}^{\infty} \frac{1}{2\pi j} \oint_{C_1} x(v) \frac{v^n}{v} dv \cdot y[n] z^{-n}$$

$$= \frac{1}{2\pi j} \oint_{C_1} \frac{x(v)}{v} dv \sum_{n=-\infty}^{\infty} y[n] v^{n-n} z^{-n}$$

$$= \frac{1}{2\pi j} \oint_{C_1} \frac{x(v)}{v} dv \sum_{n=-\infty}^{\infty} y[n] \left(\frac{z}{v}\right)^{-n}$$

$$Z \{ x[n] y[n] \} = \frac{1}{2\pi j} \oint_{C_1} x(v) y\left(\frac{z}{v}\right) v^{-1} dv$$

If C_1 is a unit circle, above equation can be written in a different form, so that it looks more like a convolution.

$$\text{Let } v = e^{j\theta} \text{ \& } z = e^{j\Omega}$$

$$dv = j e^{j\theta} d\theta$$

$$z = \{x[n]y[n]\} = \frac{1}{2\pi j} \oint_C x(e^{j\theta}) y(e^{j(\Omega-\theta)}) e^{j\theta} d\theta$$

$$= \frac{1}{2\pi} \int_{-\pi}^{\pi} x(e^{j\theta}) y(e^{j(\Omega-\theta)}) d\theta$$

denoting $x(e^{j\theta})$ as $x(\theta)$ and $y(e^{j(\Omega-\theta)})$ as $y(\Omega-\theta)$, we get,

$$z = \{x[n]y[n]\} = \frac{1}{2\pi} \int_{-\pi}^{\pi} x(\theta) y(\Omega-\theta) d\theta$$

(i) Initial Value Theorem:-

If $x[n]$ is a causal sequence with z transform $X(z)$, then $x[0] = \lim_{z \rightarrow \infty} X(z)$

$$X(z) = \sum_{n=-\infty}^{\infty} x[n] z^{-n}$$

Since $x[n]$ is a causal sequence, the lower limit in the above summation is taken as zero.

$$\text{Hence } X(z) = \sum_{n=0}^{\infty} x[n] z^{-n}$$

Taking limit $z \rightarrow \infty$ on both sides

$$\lim_{z \rightarrow \infty} X(z) = x[0] + 0$$

$$x[0] = \lim_{z \rightarrow \infty} X(z)$$

ii) Final Value Theorem:

If $Z \{x(n)\} = X(z)$ and the poles of $X(z)$ are all inside the unit circle, then final value of $x[n]$ as $n \rightarrow \infty$ given by

$$\lim_{n \rightarrow \infty} x[n] = \lim_{z \rightarrow 1} [(z-1)X(z)]$$

from the sequence,

$x(n+1) - x(n)$ and take Z transform,

$$Z \{x(n+1) - x(n)\} = \sum_{n=-\infty}^{\infty} [x(n+1) - x(n)] z^{-n}$$

The sequence $x[n]$ is causal, hence

$$Z \{x(n+1) - x(n)\} = \sum_{n=0}^{\infty} [x(n+1) - x(n)] z^{-n}$$

$$= \lim_{n \rightarrow \infty} \left[\sum_{k=0}^{\infty} [x[k+1] - x[k]] z^{-k} \right]$$

$$= \lim_{n \rightarrow \infty} \left[\sum_{k=0}^{\infty} x[k+1] z^{-k} - \sum_{k=0}^n x[k] z^{-k} \right]$$

$$= \lim_{n \rightarrow \infty} \left[-x[0] + x[1](1-z^{-1}) + (x[2](z^{-1} - z^{-2}) + \dots + x[n+1]z^{-n}) \right]$$

Taking limit $z \rightarrow 1$ on both sides we get-

$$\begin{aligned} \lim_{z \rightarrow 1} Z \{ [x(n+1)] - x(n) \} &= Z \{ x(n+1) \} - Z \{ x(n) \} \\ &= zX(z) - z(x(0) - X(z)) \\ &= (z-1)X(z) - zX(0) \end{aligned}$$

Taking $\lim_{z \rightarrow 1}$

$$\lim_{z \rightarrow 1} Z \{ x(n+1) - x(n) \} = \lim_{z \rightarrow 1} [(z-1)X(z) - zX(0)]$$

Equating above equations,

$$-x[0] + \lim_{n \rightarrow \infty} x[n] = -x[0] + \lim_{z \rightarrow 1} (z-1)X(z)$$

$$\text{Hence, } \lim_{n \rightarrow \infty} x[n] = \lim_{z \rightarrow 1} (z-1)X(z)$$

10 b. Given $X(z) = \frac{1 + 5z^{-1}}{1 - 3/2 z^{-1} + 1/2 z^{-2}}$

$$X(z) = \frac{z(5-z)}{z^2 - 3/2 z + 1/2}$$

$$\frac{X(z)}{z} = \frac{(5-z)}{(z-1/2)(z-1)}$$

$$\frac{5-z}{(z-1/2)(z-1)} = \frac{A}{(z-1/2)} + \frac{B}{(z-1)}$$

$$(5-z) = A(z-1) + B(z-1/2)$$

for $z = 1/2$

$$\frac{9}{2} = A(-1/2)$$

$$A = -9 //$$

for $z = 1$

$$4 = B(1/2)$$

$$B = 8 //$$

$$\frac{X(z)}{z} = \frac{-9}{z-1/2} + \frac{8}{z-1}$$

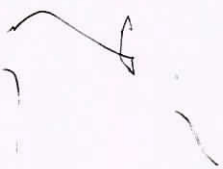
$$X(z) = \frac{-9z}{(z-1/2)} + \frac{8z}{(z-1)}$$

$$X(z) = \frac{-9}{1-1/2 z^{-1}} + \frac{8}{1-z^{-1}}$$

for ROC: $|z| > 1$ $x(n) = -9(1/2)^n u(n) + 8(1)^n u(n)$
 $= [-9(1/2)^n + 8] u(n)$

~~Q~~

```
10 c. # Z Transform
k = [2, 4, 3, 7, 5] # Discrete Sequence
ztop = 0
f = 0
for t in range(5):
    # print(k[t])
    ztop = ztop + k[t] * 2 ** -t
print(ztop)
```



~~9~~